# **Bilad Alrafidain University College Electric Power Techniques Engineering Department Control Systems Analysis Fourth Stage Academic Year 2020 - 2021**

**Lecture Seven** 

**Stability Analysis** 

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Stability of Control Systems

Stability of closed-loop feedback systems is central to control system design.

A control system is in equilibrium if:

- The absence of any disturbance or input.
- The output stays in the same state.

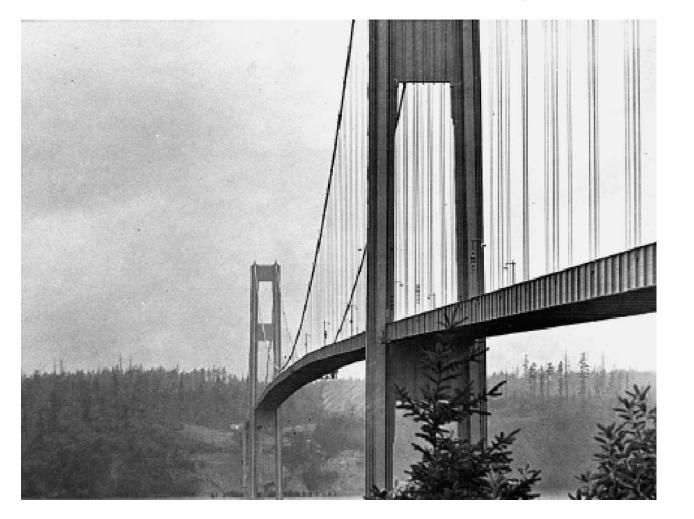
يعد إستقرار أنظمة التغذية العكسية ذات المسار المغلق أمراً أساسياً للتحكم في تصميم النظام. يكون نظام التحكم في حالة توازن إذا: • عدم وجود أي إز عاج أو إدخال للنظام. • الإخراج من النظام يبقى في نفس الحالة.

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# Stability of Control Systems

The stability of a cone: (a) Stable (b) Neutral (c) Unstable Stable Neutral Unstable

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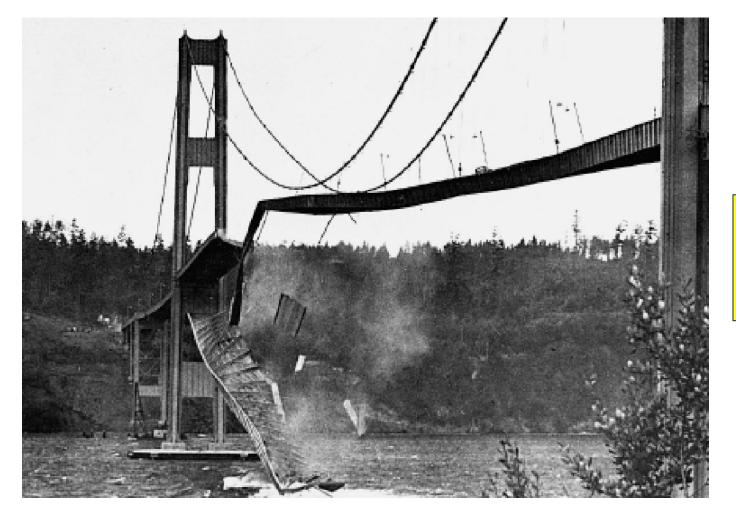


#### Stability of Control Systems

The original Tacoma Narrows Bridge was built in 1939 in the Washington State, U.S.A. At the time, it was the third longest suspension bridge in the world, and cost about US\$6.56 million (considered a bargain then). The Tacoma Narrows Bridge opened to the public on July 1, 1940.

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## Stability of Control Systems



Opened on July 1, 1940. On November 7, 1940, a windstorm induced severe torsional movement in the bridge that eventually caused the bridge to break apart and collapse.

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#### Stability of Control Systems

1. Stable

A linear time-invariant control system is stable if the output eventually comes back to its equilibrium state when the system is subjected to an initial condition.

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Stability of Control Systems

2. Critically Stable

A linear time-invariant control system is critically stable if oscillations of the output continue forever.

يكون نظام التحكم الخطي غير المتغير مع الزمن مستقرًا بشكل حرج إذا استمرت تذبذبات الإخراج إلى الأبد.

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### Stability of Control Systems

3. Unstable

A linear time-invariant control system is unstable if the output diverges without bound from its equilibrium state when the system is subjected to an initial condition.

يكون نظام التحكم الخطي غير المتغير مع الزمن غير مستقراً إذا تباعد الإخراج دون تقييد من حالة توازنه عندما يخضع النظام لشرط مبدئي.

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# Stability of Control Systems

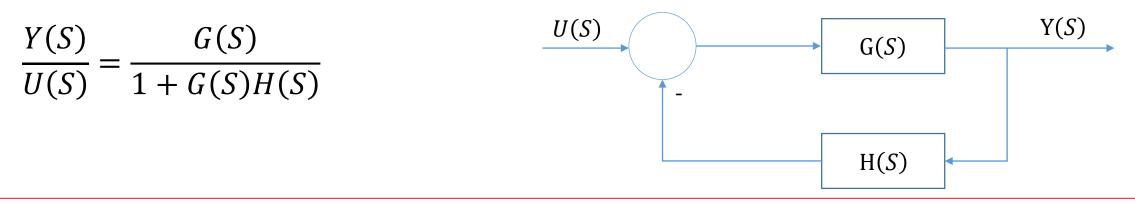
#### Note:

Actually, the output of a physical system may increase to a certain extent but may be limited by mechanical "stops," or the system may break down or become nonlinear after the output exceeds a certain magnitude so that the linear differential equations no longer apply.

ملاحظة في الواقع ، قد يزداد إخراج النظام المادي إلى حد معين ولكن قد يكون محدوداً بواسطة "نقاط توقف ميكانيكة"، أو قد ينهار النظام أو يصبح غير خطي بعد أن يتجاوز إخراج النظام قيمة معينة بحيث لا يتم تطبيق المعادلات التفاضلية الخطية

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## Stability Methods for Control Systems

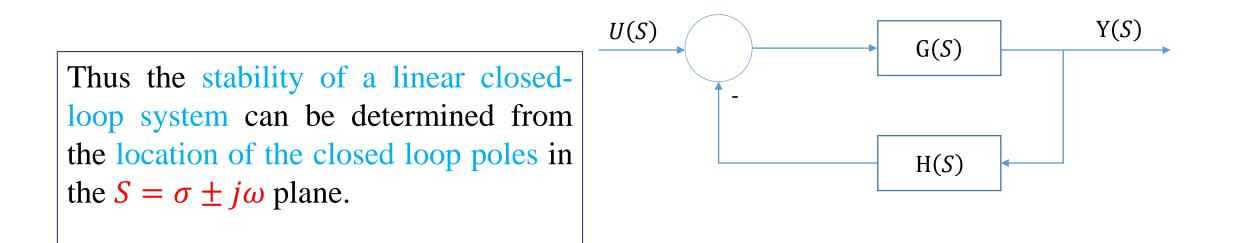


- Solving *characteristic equation*, P(s) = 0 or matrix P(s) = |sI A| = 0, to determine the stability based on the location of roots on the complex  $S = \sigma \pm j\omega$  plane.
- Applying *Routh-Hurwitz stability criterion* (*Routh* 1872, *Hurwitz* 1892), without solving P(s) = 0.
- Applying *Root Locus* (*Evans* 1948) Frequency analysis methods.
- Bode diagram (Bode 1927),  $P(j\omega) = 0$ .
- *Nyquist stability criterion* (1932), without solving  $P(j\omega) = 0$ .
- Lyapunov stability methods, (Lyapunov 1892) (become known in west 1961)

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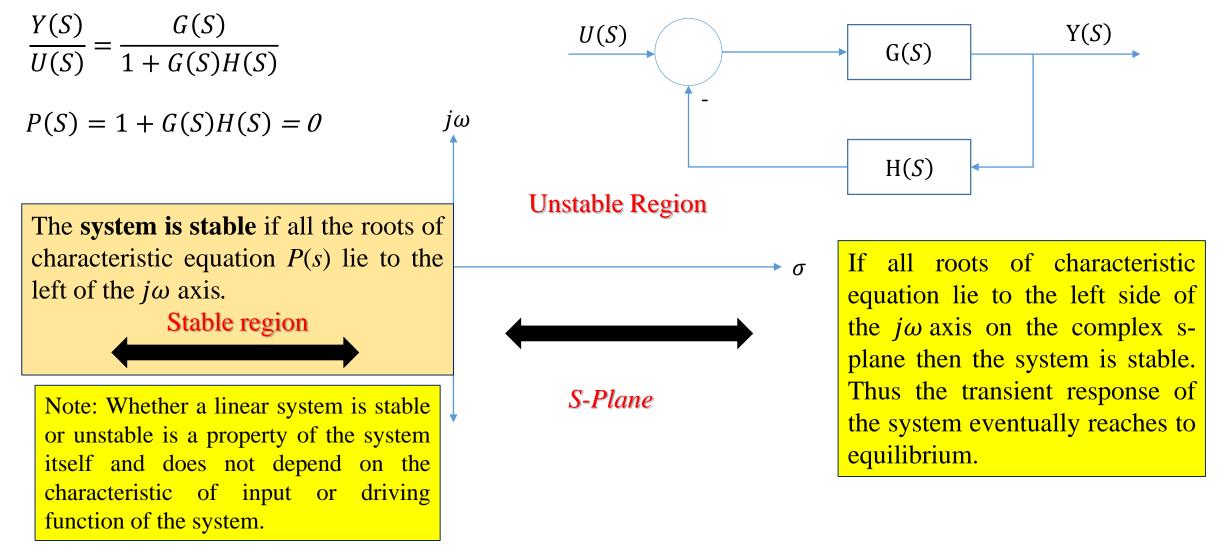
Stability of Linear Feedback Systems by Solving Characteristic Equation

The stability of a feedback system is directly related to the location of the roots of the characteristic equation of the system transfer function and to the location of the eigenvalues of the system matrix for a system in state variable format.

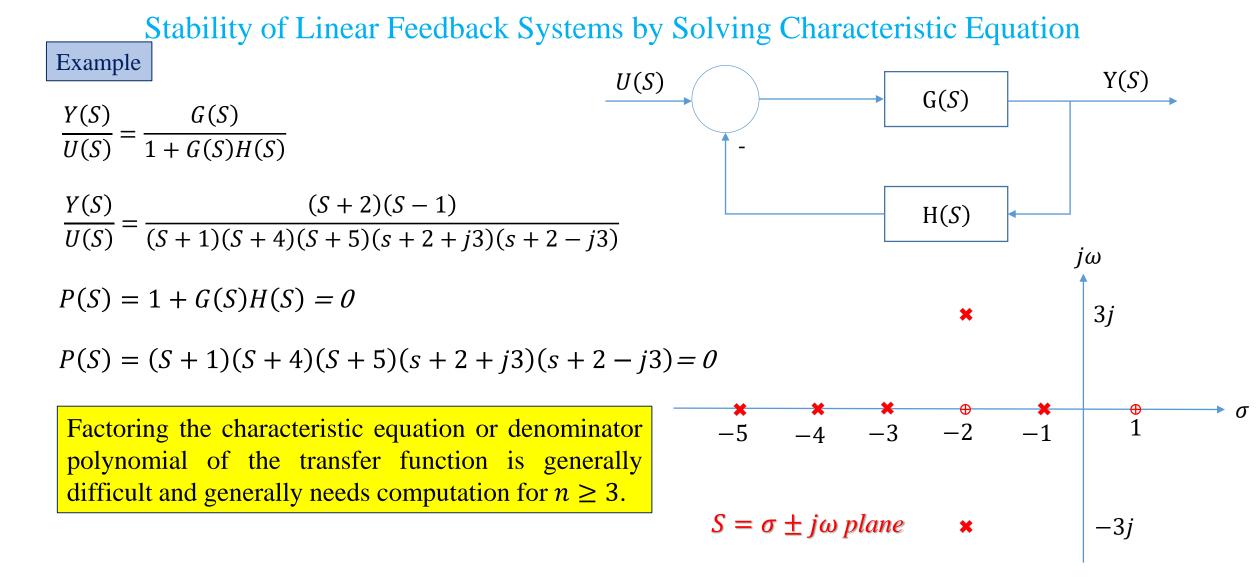


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## Stability of Linear Feedback Systems by Solving Characteristic Equation



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Stability of Linear Feedback Systems by Solving Characteristic Equation Y(S)U(S) $\frac{Y(S)}{U(S)} = \frac{G(S)}{1 + G(S)H(S)}$ G(S)H(S)State Space Model **Transfer Function Model**  $\frac{Y(S)}{U(S)} = \frac{G(S)}{1 + G(S)H(S)}$  $\dot{x} = Ax + Bu$ y = Cx + Du $C \in R^{m*n}$  $x \in \mathbb{R}^n$  $D \in R^{m*r}$  $u \in R^r$  $A \in R^{n*n}$ P(s) = 1 + G(S)H(S) = 0 $B \in R^{n*r}$ P(s) = |sI - A| = 0